

A Wide-Band Parametric Amplifier with Triple-Tuned Signal Circuit and Single-Tuned Idler Circuit

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It is the purpose of this letter to present the results of the extension of the work of Egami to the case of a single-tuned idler circuit and a triple-tuned signal circuit. As shown by Egami [1], the input impedance of the double-tuned amplifier is given by

$$Z' = \frac{\left[R_s + jX_1\Omega - \frac{1}{1/R_a + jb_a\Omega} \right]}{1 + jb_b\Omega \left[R_s + jX_1\Omega - \frac{1}{1/R_a + jb_a\Omega} \right]} \quad (1)$$

$$\Gamma(S) = \frac{\alpha(1+S)(1+\tilde{a}S)(1-\beta\eta S+S^2\tilde{\alpha}\Omega t) - (1+S)(\beta-S\tilde{\alpha}\Omega t) - (1-\beta\eta S+\eta S^2\tilde{\alpha}\Omega t)}{\alpha(1+S)(1+\tilde{a}S)(1+\beta\eta S+S^2\eta\tilde{\alpha}\Omega t) + (1+S)(\beta+S\tilde{\alpha}\Omega t) - (1+\beta\eta S+\eta S^2\alpha\tilde{\alpha}\Omega t)} \quad (5)$$

If a series-tuned circuit resonant at the signal frequency is connected across the input terminal Fig. 1, the input impedance is given by

$$Z = \frac{R_s + jX_1\Omega - \frac{1}{1/R_a + jb_a\Omega}}{1 + jb_b\Omega \left[R_s + jX_1\Omega - \frac{1}{1/R_a + jb_a\Omega} \right]} + jX_3\Omega \quad (2)$$

$$Z = \frac{R_a \left[\alpha(1+S\tilde{a}) - \frac{1}{(1+S)} \right] + jX_3\Omega \left[1 + \eta S \left\{ \alpha(1+S\tilde{a}) - \frac{1}{(1+S)} \right\} \right]}{1 + \eta S \left[\alpha(1+S\tilde{a}) - \frac{1}{(1+S)} \right]} \quad (3)$$

where

$$\Omega = \frac{\omega_1}{\omega_{10}} - \frac{\omega_{10}}{\omega} \simeq \frac{2\Delta\omega_1}{\omega_{10}}, \quad \Delta\omega_1 = \omega_1 - \omega_{10}$$

$$\Omega_0 = \frac{R_s}{a_2 x_{10}} (\tilde{Q}_1 \cdot \tilde{Q}_2 - 1)^{1/2}$$

$$\eta = b_b/b_a, \quad t = x_b/x_1$$

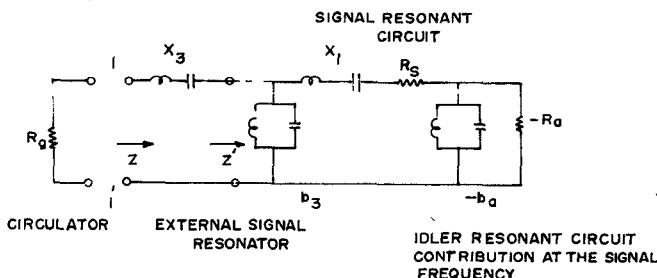


Fig. 1. Signal frequency equivalent circuit of wide-band paramp.

Manuscript received May 28, 1975; revised September 30, 1975.

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$$\alpha = R_s/R_a = \frac{1}{\tilde{Q}_1 \tilde{Q}_2}, \quad \beta = R_b/R_a$$

$$\tilde{a} = a_1/a_2$$

$$S = ja_2 \frac{x_{10}}{R_s} \Omega$$

$$R_a = \tilde{Q}_1 \tilde{Q}_2 R_s$$

$$b_a = \frac{a_2 x_{10}}{R_a R_s}.$$

The reflection coefficient at 1 - 1' port in Fig. 1 is given by

$$\Gamma(S) = \frac{Z - R_b}{Z + R_b}. \quad (4)$$

Thus

The gain $|\Gamma(S)|^2$ has been calculated with the help of a digital computer and is plotted for the various values of the different parameters. From the results it is observed that the optimum value of slope parameter t of the third-signal resonator was found to be 0.315 for which the ripple in passband is minimum.

As an example, gain $|\Gamma(S)|^2$ is shown in Fig. 2 as a function of Ω/Ω_0 , with $\alpha = 0.05$ ($\tilde{Q}_1 \cdot \tilde{Q}_2 = 20$), $a_1/a_2 = 1.0$, $\beta = 1.829$, $t = 0.315$, and varying η . The dotted curves shown in Fig. 2 are for a single-tuned idler and a double-tuned signal while the solid curves are for a single-tuned idler and a triple-tuned signal. As seen from the

curves in Fig. 2, the passband ripple and its bandwidth appear to be affected by using a triple-tuned stage, more than by using a double-tuned signal stage. It is apparent from Fig. 2, that the ripple has been decreased whereas the gain remains the same.

Fig. 3 represents the solution corresponding to 10-dB gain, ± 1 -dB ripple, $a_1/a_2 = 1.0$, and three values of α (or $\tilde{Q}_1 \cdot \tilde{Q}_2$) for double- and triple-tuned signal. This graph shows that a large value of $\tilde{Q}_1 \cdot \tilde{Q}_2$ does not increase the bandwidth ratio Ω/Ω_0 in the triple-tuned case, but the gain increases and the ripple decreases in the passband for the triple-tuned signal when $\alpha = 0.05$ compared to the double-tuned case.

Fig. 4 represents the solution corresponding to 10-dB gain, ± 1 -dB ripple, $\tilde{Q}_1 \cdot \tilde{Q}_2 = 20.0$, and four values of a_1/a_2 for the double- and triple-tuned signal and the single-tuned idler. From the graph, it is apparent that for $a_1/a_2 = 1.0$, the gain is high and ripple is low for the triple-tuned signal circuit compared to the double-tuned signal circuit.

ACKNOWLEDGMENT

The authors wish to thank Dr. A. K. Kamal, Fellow, IEEE, Professor, and Head of the Department for discussion and encouragement.

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- [1] S. Egami, "A design theory for wide-band parametric amplifiers," *IEEE Trans. Microwave Theory Tech.*, vol. MTT-22, pp. 119-124, Feb. 1974.

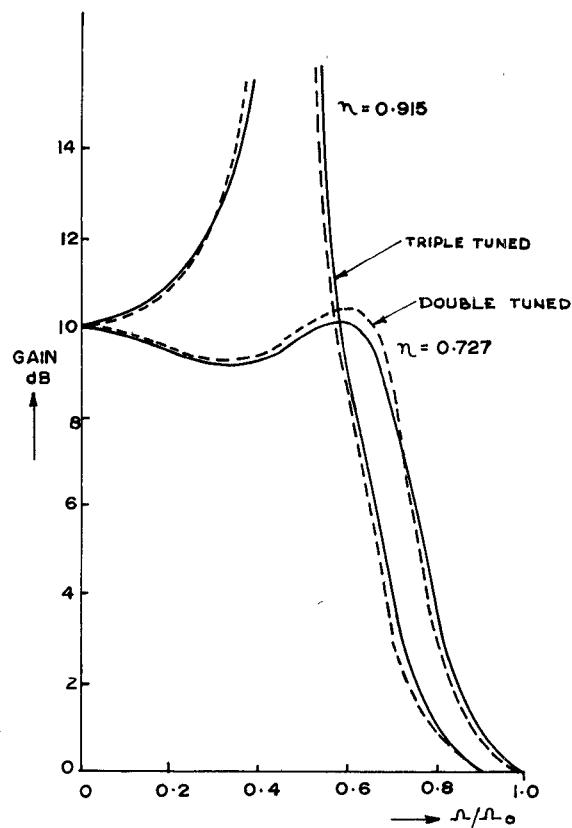


Fig. 2. Theoretical gain frequency response varying the ratio of external signal resonator slope parameter to the negative slope parameter of the pumped diode. —: triple-tuned signal. - - -: double-tuned signal.

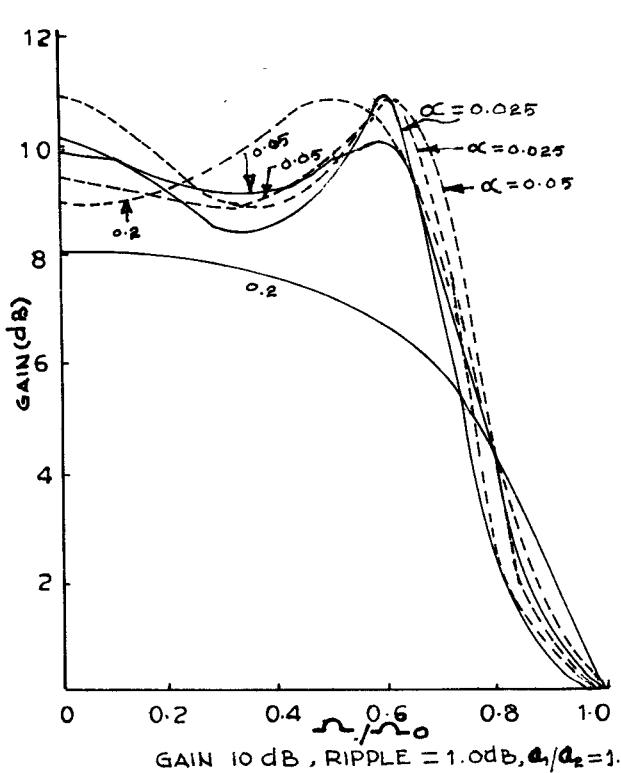


Fig. 3. Theoretical gain frequency response for different α . —: triple-tuned signal. - - -: double-tuned signal.

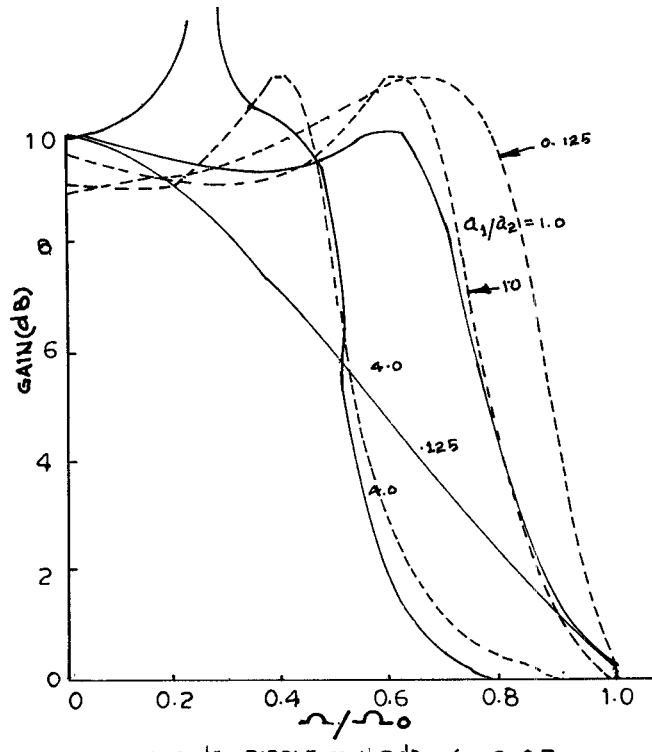


Fig. 4. Theoretical gain frequency response for different a_1/a_2 . —: triple-tuned signal. - - -: double-tuned signal.